# **LM386**

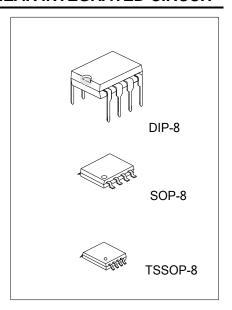
# LINEAR INTEGRATED CIRCUIT

# LOW VOLTAGE AUDIO POWER AMPLIFIER

#### ■ DESCRIPTION

The UTC **LM386** is a power amplifier, designed for use in low voltage consumer applications. The gain is internally set to 20 to keep external part count low, but the addition of an external resistor and capacitor between pin 1 and pin 8 will increase the gain to any value up from 20 to 200.

The inputs are ground referenced while the output automatically biases to one-half the supply voltage. The quiescent power drain is only 24 milliwatts when operating from a 6 volt supply, making the **LM386** ideal for battery operation.

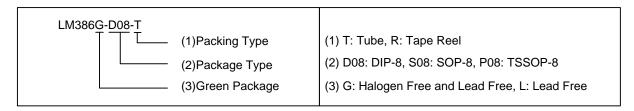


#### **■ FEATURES**

- \*Battery Operation
- \*Minimum External Parts
- \*Wide Supply Voltage Range: 4V~12V \*Low Quiescent Current Drain:4mA
- \*Voltage Gains: 20~200
  \*Ground Referenced Input
- \*Self-Centering Output Quiescent Voltage
- \*Low Distortion: 0.2% (Av=20,  $V_S$ =6V,  $R_L$ =8 $\Omega$ ,  $P_O$ =125mW, f=1kHz)

#### **■ ORDERING INFORMATION**

Ordering Number		Dookogo	Packing	
Lead Free	Halogen Free	Halogen Free Package		
LM386L-D08-T	LM386G-D08-T	DIP-8	Tube	
LM386L-S08-R	LM386G-S08-R	SOP-8	Tape Reel	
LM386L-P08-R	LM386G-P08-R	TSSOP-8	Tape Reel	

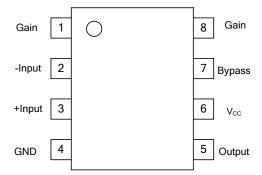


www.unisonic.com.tw 1 of 7

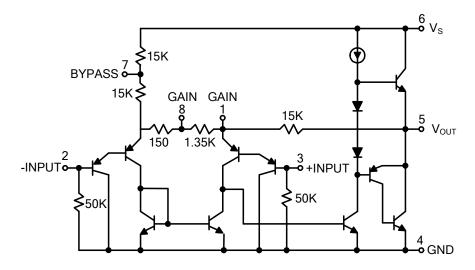
#### ■ MARKING

PACKAGE	MARKING		
DIP-8	Date Code  UTC DDD L: Lead Free  LM386 G: Halogen Free  Lot Code		
SOP-8	B 7 6 5  UTC DDDD LM386 L: Lead Free  LM386 Lot Code  1 2 3 4 Lot Code		
TSSOP-8	Date Code LM386  Li Lead Free G: Halogen F Lot Code		

## **■ PIN CONFIGURATION**



# ■ BLOCK DIAGRAM



#### ■ ABSOLUTE MAXIMUM RATINGS

PARAMET	ER	SYMBOL	RATINGS	UNIT
Supply Voltage		V <sub>cc</sub>	15	V
Input Voltage		V <sub>IN</sub>	-0.4V ~ +0.4V	V
Power Dissipation	DIP-8		1250	mW
	SOP-8	P <sub>D</sub>	600	mW
	TSSOP-8		600	mW
Junction Temperature		TJ	+125	°C
Operating Temperature		T <sub>OPR</sub>	-40 ~ <b>+</b> 85	°C
Storage Temperature	_	T <sub>STG</sub>	-40 ~ <b>+</b> 150	°C

Note:1. Absolute maximum ratings are stress ratings only and functional device operation is not implied. The device could be damaged beyond Absolute maximum ratings.

## ■ **ELECTRICAL CHARACTERISTICS** (T<sub>A</sub>=25°C, unless otherwise specified.)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Operating Supply Voltage	Vs		4		12	V	
Quiescent Current	$I_Q$	$V_S=6V, V_{IN}=0$		4	8	mΑ	
Output Power	Pout	$V_S=6V$ , $R_L=8\Omega$ , $THD=10\%$	250	325		2014/	
		$V_S=9V$ , $R_L=8\Omega$ , $THD=10\%$	500	700		mW	
Voltage Gain	G∨	V <sub>S</sub> =6V, f=1kHz		26		dB	
		10µF from pin 1 to pin 8		46		dB	
Bandwidth	BW	V <sub>S</sub> =6V , Pin1 and pin 8 open		300		kHz	
Total Harmonic Distortion	I IHI)	P <sub>OUT</sub> =125mW, V <sub>S</sub> =6V, f=1kHz	0.2		%		
		R <sub>L</sub> =8Ω pin1 and pin 8 open	0.2				
Rejection Ratio	RR	V <sub>S</sub> =6V, f=1kHz, C <sub>BYPASS</sub> =10µF	50			dB	
		pin1and pin 8 open, Referred to output				иБ	
Input Resistance	R <sub>IN</sub>			50		kΩ	
Input Bias Current	I <sub>BIAS</sub>	V <sub>S</sub> =6V Pin2 and pin 3 open		250		nA	

#### ■ APPLICATION NOTES

#### **GAIN CONTROL**

To make the **LM386** a more versatile amplifier, two pins(1and 8) are provided for gain control. With pins 1 and 8 open the 1.35  $k\Omega$  resistor sets the gain at 20 (26dB), If a capacitor is put from pin 1 to 8, bypassing the 1.35  $k\Omega$  resistor, the gain will go up to 200 (46dB). If a resistor is placed in series with the capacitor, the gain can be set to any value from 20 to 200. Gain control can also be done by capacitively coupling a resistor (or FET) from pin 1 to ground.

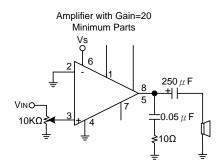
Additional external components can be placed in parallel with the internal feedback resistors to tailor the gain and frequency response for individual applications. For example we can compensate poor speaker bass response by frequency shaping the feedback path. This is done with a series RC from pin 1 to 5 (paralleling the internal 15 k $\Omega$  resistor). For 6 dB effective bass boost: R=15 k $\Omega$ , the lowest value for good stable operation is R=10 k $\Omega$ , if pin 8 is open, If pins 1 and 8 are bypassed then R as low as 2 k $\Omega$  can be used. This restriction is because the amplifier is only compensated for closed-loop gains greater than 9.

#### **INPUT BIASING**

The schematic shows that both inputs are biased to ground with a 50 k $\Omega$  resistor. The base current of the input transistors is about 250nA, so the inputs are at about 12.5mW when left open. If the dc source resistance driving the **LM386** is higher than 250 k $\Omega$  it will contribute very little additional offset (about 2.5mW at the input, 50mW at the output). If the dc source resistance is less than 10 k $\Omega$ , then shorting the unused input to ground will keep the offset low (about 2.5mV at the input, 50 mV at the output). For dc source resistances between these values we can eliminate excess offset by putting a resistor from the unused input to ground, equal in value to the dc source resistance. Of course all offset problems are eliminated if the input is capacitively coupled.

When using the **LM386** with higher gains (bypassing the 1.35 k $\Omega$  resistor between pins 1 and 8) it is necessary to bypass the unused input, preventing degradation of gain and possible instabilities. This is done with a 0.1 $\mu$ F capacitor or a short to ground depending on the dc source resistance on the driven input.

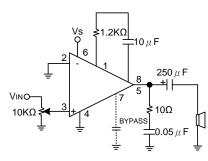
#### **■ TYPICAL APPLICATIONS CIRCUIT**

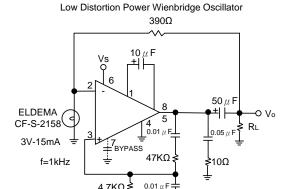


 $\begin{array}{c|c}
 & 10 \mu \text{F} \\
\hline
2 & 10 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text{K} \Omega & 250 \mu \text{F} \\
\hline
10 \text$ 

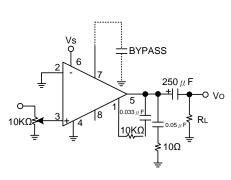
Amplifier with Gain=200

Amplifier with Gain=50

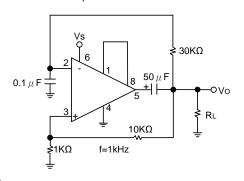




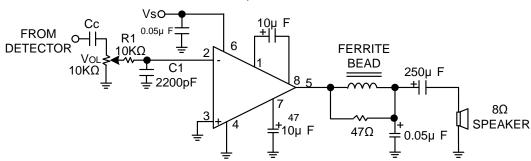
Amplifier with Bass Boost



Square Ware Oscillator

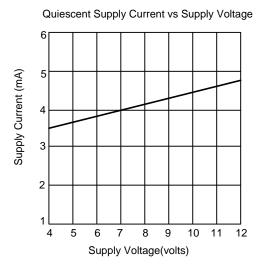


AM Radio Power Amplifier

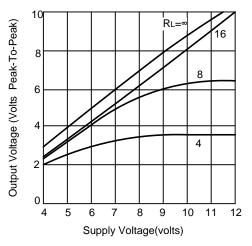


- Notes: 1: Twist Supply lead and supply ground very tightly.
  - 2: Twist speaker lead and ground very tightly.
  - 3: Ferrite bead in Ferroxcube K5-001-001/3B with 3 turns of wire.
  - 4: R1C1 band limits input signals.
  - 5: All components must be spaced very closely to IC.

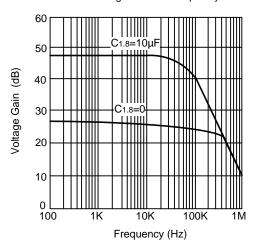
#### **■ TYPICAL CHARACTERISTICS**



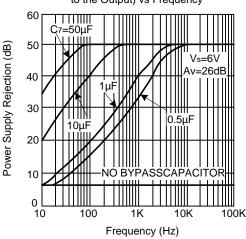
Peak-to -Peak Output Voltage Swing vs Supply Voltage



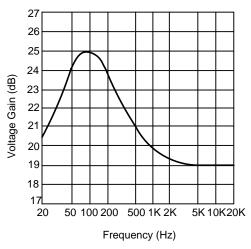
Voltage Gain vs Frequency



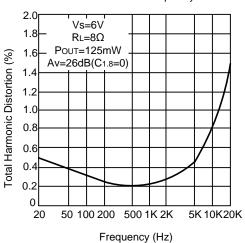
Power Supply Rejection Ratio(Referred to the Output) vs Frequency



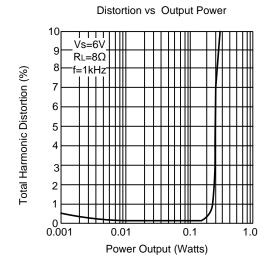
Frequency Response With Bass Boost

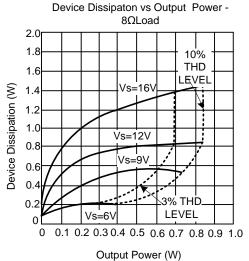


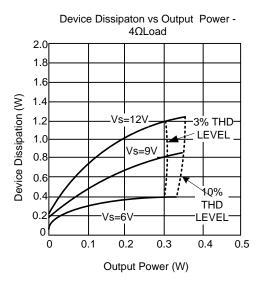
Distortion vs Frequency

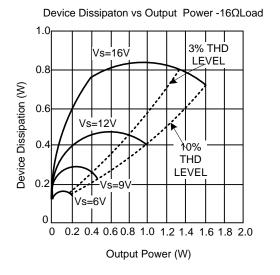


# **■ TYPICAL CHARACTERISTICS (Cont.)**









UTC assumes no responsibility for equipment failures that result from using products at values that exceed, even momentarily, rated values (such as maximum ratings, operating condition ranges, or other parameters) listed in products specifications of any and all UTC products described or contained herein. UTC products are not designed for use in life support appliances, devices or systems where malfunction of these products can be reasonably expected to result in personal injury. Reproduction in whole or in part is prohibited without the prior written consent of the copyright owner. UTC reserves the right to make changes to information published in this document, including without limitation specifications and product descriptions, at any time and without notice. This document supersedes and replaces all information supplied prior to the publication hereof.